

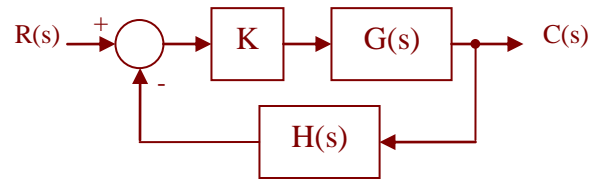
Rules for Making Root Locus Plots

The closed loop transfer function of the system shown is

$$T(s) = \frac{KG(s)}{1 + KG(s)H(s)}$$

So the characteristic equation (*c.e.*) is

$$1 + KG(s)H(s) = 1 + K \frac{N(s)}{D(s)} = 0, \text{ or } D(s) + KN(s) = 0.$$



As K changes, so do locations of closed loop poles (i.e., zeros of *c.e.*). The table below gives rules for sketching the location of these poles for $K=0 \rightarrow \infty$ (i.e., $K \geq 0$).

| Rule Name | Description |
|---|--|
| Definitions | <ul style="list-style-type: none"> The loop gain is $KG(s)H(s)$ or $K \frac{N(s)}{D(s)}$. $N(s)$, the numerator, is an m^{th} order polynomial; $D(s)$, is n^{th} order. $N(s)$ has zeros at z_i ($i=1..m$); $D(s)$ has them at p_i ($i=1..n$). The difference between n and m is q, so $q=n-m$. ($q \geq 0$) |
| Symmetry | The locus is symmetric about real axis (i.e., complex poles appear as conjugate pairs). |
| Number of Branches | There are n branches of the locus, one for each closed loop pole. |
| Starting and Ending Points | The locus starts ($K=0$) at poles of loop gain, and ends ($K \rightarrow \infty$) at zeros. Note: this means that there will be q roots that will go to infinity as $K \rightarrow \infty$. |
| Locus on Real Axis* | The locus exists on real axis to the left of an odd number of poles and zeros. |
| Asymptotes as $s \rightarrow \infty$* | If $q > 0$ there are asymptotes of the root locus that intersect the real axis at $\sigma = \frac{\sum_{i=1}^m p_i - \sum_{i=1}^n z_i}{q}$, and radiate out with angles $\theta = \pm r \frac{180}{q}$, where $r=1, 3, 5 \dots$ |
| Break-Away/-In Points on Real Axis | Break-away or -in points of the locus exist where $N(s)D'(s) - N'(s)D(s) = 0$. |
| Angle of Departure from Complex Pole* | Angle of departure from pole, p_j is $\theta_{\text{depart}, p_j} = 180^\circ + \sum_{i=1}^m \angle(p_j - z_i) - \sum_{\substack{i=1 \\ i \neq j}}^n \angle(p_j - p_i)$. |
| Angle of Arrival at Complex Zero* | Angle of arrival at zero, z_j , is $\theta_{\text{arrive}, z_j} = 180^\circ - \sum_{i=1}^m \angle(z_j - p_i) + \sum_{\substack{i=1 \\ i \neq j}}^n \angle(z_j - z_i)$. |
| Locus Crosses Imaginary Axis | Use Routh-Hurwitz to determine where the locus crosses the imaginary axis. |
| Given Gain "K," Find Poles | Rewrite <i>c.e.</i> as $D(s) + KN(s) = 0$. Put value of K into equation, and find roots of <i>c.e.</i> (This may require a computer) |
| Given Pole, Find "K." | Rewrite <i>c.e.</i> as $K = -\frac{D(s)}{N(s)}$, replace "s" by desired pole location and solve for K . Note: if "s" is not exactly on locus, K may be complex (small imaginary part). Use real part of K . |

*These rules change to draw complementary root locus ($K \leq 0$). See next page for details.

Complementary Root Locus

To sketch complementary root locus ($K \leq 0$), most of the rules are unchanged except for those in table below.

| Rule Name | Description |
|--|--|
| Locus on Real Axis | The locus exists on real axis to the right of an odd number of poles and zeros. |
| Asymptotes as $s \rightarrow \infty$ | If $q > 0$ there are asymptotes of the root locus that intersect the real axis at $\sigma = \frac{\sum_{i=1}^m p_i - \sum_{i=1}^n z_i}{q}$, and radiate out with angles $\theta = \pm p \frac{180}{q}$, where $p=0, 2, 4, \dots$ |
| Angle of Departure from Complex Pole | Angle of departure from pole, p_j is $\theta_{\text{depart}, p_j} = \sum_{i=1}^m \angle(p_j - z_i) - \sum_{\substack{i=1 \\ i \neq j}}^n \angle(p_j - p_i).$ |
| Angle of Departure at Complex Zero | Angle of arrival at zero, z_j , is $\theta_{\text{arrive}, z_j} = \sum_{\substack{i=1 \\ i \neq j}}^m \angle(z_j - z_i) - \sum_{i=1}^n \angle(z_j - p_i).$ |

Other Forms of Root Locus

Not yet complete.

Design with Root Locus

Not yet complete.